

600mA、1.5MHz、同步降压型 DC-DC 转换器**UM3500 SOT23-5****描述**

UM3500 是一款 1.5MHz 恒频、采用斜率补偿电流模式的 PWM 降压转换器。该器件内置了一个主开关和一个同步整流器，无需外部肖特基二极管即可实现高效率。该器件适用于为单节锂离子电池的便携式设备供电。UM3500 可从 2.5V 至 5.5V 输入电压范围内提供 600mA 的负载电流。输出电压最低可调至 0.6V。UM3500 还支持 100% 占空比的低压差运行模式，可延长便携式系统的电池寿命。在轻载条件下，该器件采用脉冲跳跃模式运行，可提供极低的输出纹波电压，适用于噪声敏感型应用。

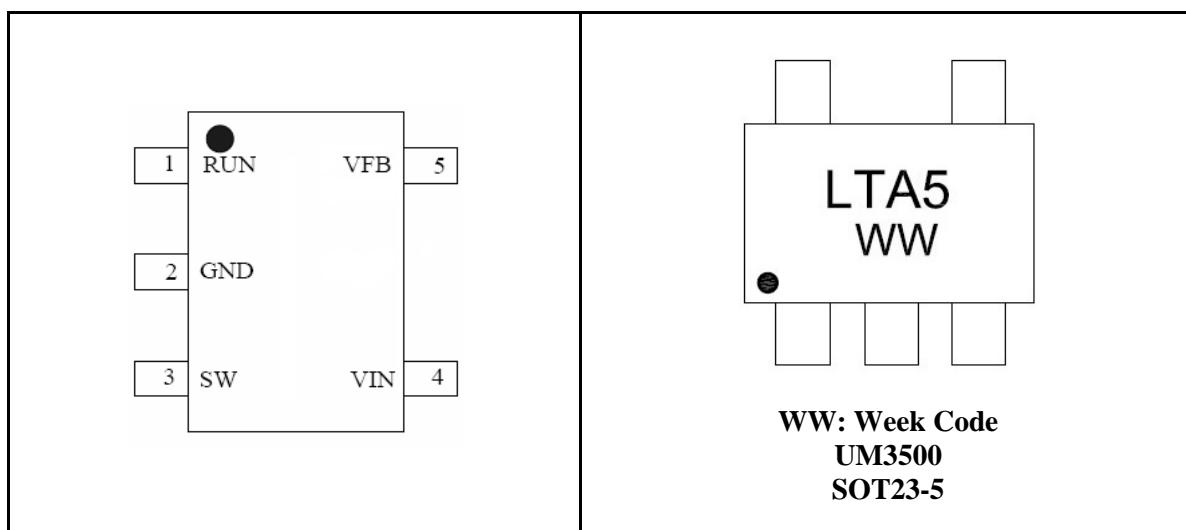
UM3500 采用扁平 SOT23-5 封装。

应用

- 蜂窝电话和智能手机
- 微处理器和 DSP 核心电源
- 无线和 DSL 调制解调器
- PDA、GPS
- MP3 播放器
- 便携式仪器

特性

- 效率高达 96%
- 1.5MHz 恒定开关频率
- 600mA 输出电流
- 内置主开关和同步整流器。无需外接肖特基二极管
- 输入电压范围：2.5V 至 5.5V
- 输出电压低至 0.6V
- 低压差运行：100% 占空比
- 低静态电流：50µA
- 短路保护
- 过温保护
- 关断电流 <1µA
- 无铅 SOT23-5 封装

引脚配置**顶部视图**

Ordering Information

| Part Number | Packaging Type | Marking Code | Shipping Qty |
|-------------|----------------|--------------|------------------------------|
| UM3500 | SOT23-5 | LTA5 | 3000pcs/7Inch Tape & Reel |

Pin Description

| Pin Number | Symbol | Function |
|------------|--------|--|
| 1 | RUN | Regulator enables control input. Drive RUN above 1.0V to turn on the part. Drive RUN below 0.4V to turn it off. In shutdown, all functions are disabled drawing $<1\mu A$ supply current. Do not leave RUN floating. |
| 2 | GND | Ground. |
| 3 | SW | Power switch output. It is the switch node connection to Inductor. This pin connects to the drains of the internal P-CH and N-CH MOSFET switches. |
| 4 | VIN | Supply input pin. Must be closely decoupled to GND, Pin 2, with a $2.2\mu F$ or greater ceramic capacitor. |
| 5 | VFB | Feedback input pin. Connect FB to the center point of the external resistor divider. Bonding option |

Absolute Maximum Ratings (Note 1)

| Symbol | Parameter | Value | Unit |
|-------------------|---------------------------------|----------------------|------|
| V_{IN} | Input Voltage | -0.3 to +6.0 | V |
| V_{RUN}, V_{FB} | RUN, VFB Voltages | +0.3 to $V_{IN}+0.3$ | V |
| V_{SW}, V_{OUT} | SW, VOUT Voltages | +0.3 to $V_{IN}+0.3$ | V |
| I_{SW} | Peak SW Sink and Source Current | 1.5 | A |
| T_O | Operating Temperature | -40 to +85 | °C |
| T_{STG} | Storage Temperature Range | -65 to +150 | °C |

Note 1: Stresses greater than those listed under Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

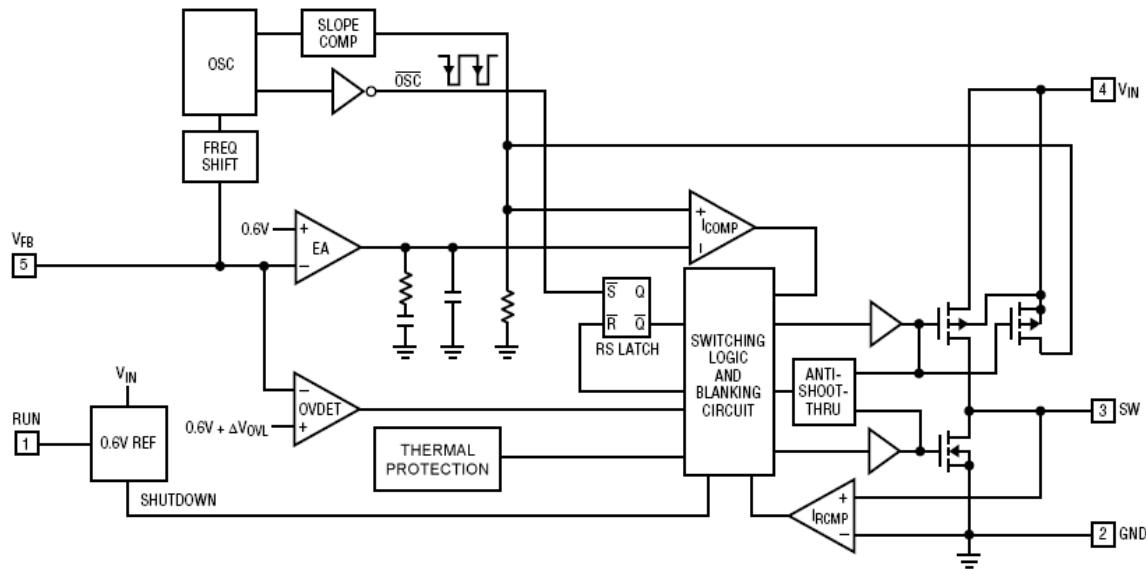
Electrical Characteristics (Note 2)

(V_{IN}=V_{RUN}=3.6V, T_A=+25 °C, unless otherwise noted)

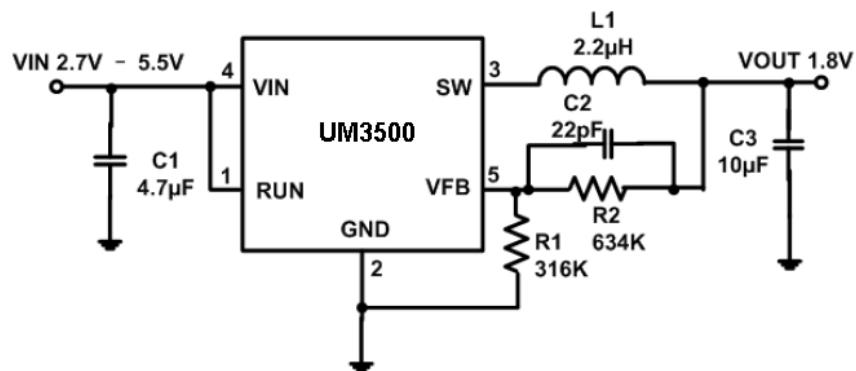
| Symbol | Parameter | Test Conditions | Min | Typ | Max | Unit |
|------------------------------|--|--|------------|------------|------------|-------------|
| V _{IN} | Input Voltage Range | | 2.5 | | 5.5 | V |
| I _Q (Active) | Input DC Supply Current (Active Mode) | V _{FB} =3.6V, V _{FB} =V _{REF} +5% | | 50 | 100 | µA |
| I _Q (Shutdown) | Input DC Supply Current (Shutdown Mode) | V _{FB} =0V, V _{IN} =4.2V | | 0.08 | 1.0 | µA |
| V _{FB} | Regulated Feedback Voltage | T _A =+25 °C | 0.5880 | 0.6000 | 0.6120 | V |
| | | 0 °C ≤ T _A ≤ 85 °C | 0.5865 | 0.6000 | 0.6135 | |
| | | -40 °C ≤ T _A ≤ 85 °C | 0.5850 | 0.6000 | 0.6150 | |
| I _{FB} | V _{FB} Input Bias Current | V _{FB} =0.65V | | | ±30 | nA |
| | Reference Voltage Line Regulation | 2.5V ≤ V _{IN} ≤ 5.5V, V _{OUT} =V _{FB} (R ₂ =0) | | 0.11 | 0.40 | %/V |
| | Output Voltage Line Regulation | 2.5V ≤ V _{IN} ≤ 5.5V, I _{OUT} =10mA | | 0.11 | 0.40 | %/V |
| | Output Voltage Load Regulation | 100mA ≤ I _{OUT} ≤ 600mA | | 0.0015 | | %/mA |
| I _{O(Max)} | Maximum Output Current | V _{IN} =3.6V, V _{OUT} =1.8V | 600 | | | mA |
| f | Oscillator Frequency | V _{FB} =0.6V or V _{OUT} =100% | 1.2 | 1.5 | 1.8 | MHz |
| R _{DS(ON)} | R _{DS(ON)} of P-CH MOSFET | V _{IN} =3.6V, I _{SW} =100mA | | 0.40 | 0.50 | Ω |
| | R _{DS(ON)} of N-CH MOSFET | V _{IN} =3.6, I _{SW} =-100mA | | 0.35 | 0.45 | Ω |
| I _P | Peak Inductor Current | V _{IN} =3.0V, V _{FB} =0.5V or V _{OUT} =90%, Duty Cycle<35% | | 1.20 | | A |
| I _{SWL} | SW Leakage | V _{RUN} =0V, V _{SW} =0V or 5V V _{IN} =5V | | ±0.01 | ±1 | µA |
| V _H | RUN High-Level Threshold | -40 °C ≤ T _A ≤ 85 °C | 1.0 | | | V |
| V _L | RUN Low-Level Threshold | -40 °C ≤ T _A ≤ 85 °C | | | 0.4 | V |
| I _{RUNL} | RUN Leakage Current | V _{RUN} =3.6V | | 2.0 | 3.5 | µA |
| η(MAX) | Max. Efficiency | V _{IN} =2.7V, V _{OUT} =2.5V | 90 | | | % |
| | Thermal Shutdown Temp | | | 160 | | °C |

Note2: 100% production test at +25 °C. Specifications over the temperature range are guaranteed by design and characterization.

Block Diagram



Typical Application Circuit



Function Description

UM3500 is a monolithic switching mode Step-Down DC-DC converter. It utilizes internal MOSFETs to achieve high efficiency and can generate very low output voltage by using internal reference at 0.6V. It operates at a fixed switching frequency, and uses the slope compensated current mode architecture. This Step-Down DC-DC Converter supplies 600mA output current at $V_{IN} = 3.6V$ with input voltage range from 2.5V to 5.5V.

Current Mode PWM Control

Slope compensated current mode PWM control provides stable switching and cycle by cycle current limit for excellent load and line responses. During normal operation, the internal main switch is turned on for a certain time to ramp the inductor current at each rising edge of the internal oscillator, and turned off when the peak inductor current reaches the controlled value. When the main switch is off, the synchronous rectifier will be turned on immediately and stay on until either the inductor current starts to reverse, as indicated by the current reversal comparator, IRCMP, or the beginning of the next clock cycle.

Pulse Skipping Mode Operation

At very light loads, the UM3500 will automatically enter Pulse Skipping Mode to increase efficiency, further extending battery life. In this mode, the control loop skips PWM pulses while maintaining output in regulation, and the switching frequency depends on the load condition. This is a kind of PFM mode operation.

Dropout Operation

When the input voltage decreases toward the value of the output voltage, the UM3500 allows the main switch to remain on for more than one switching cycle and increases the duty cycle (Note 1) until it reaches 100%. The output voltage then is the input voltage minus the voltage drop across the main switch and the inductor. At low input supply voltage, the $R_{DS(ON)}$ of the P-Channel MOSFET increases, and the efficiency of the converter decreases. Caution must be exercised to ensure the heat dissipated not to exceed the maximum junction temperature of the IC.

Note 1: The duty cycle D of a step-down converter is defined as:

$$D = T_{ON} \times f_{OSC} \times 100\% \cong V_{OUT}/V_{IN} \times 100\%$$

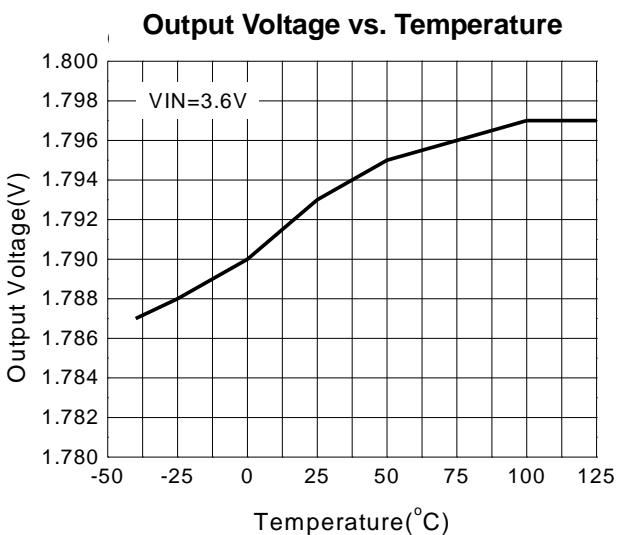
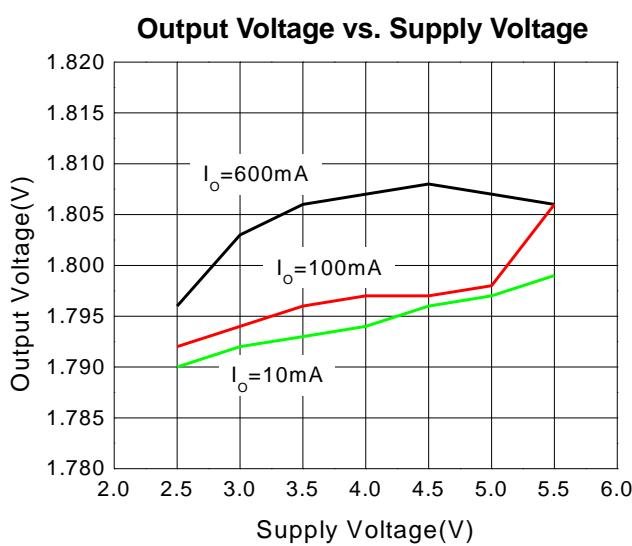
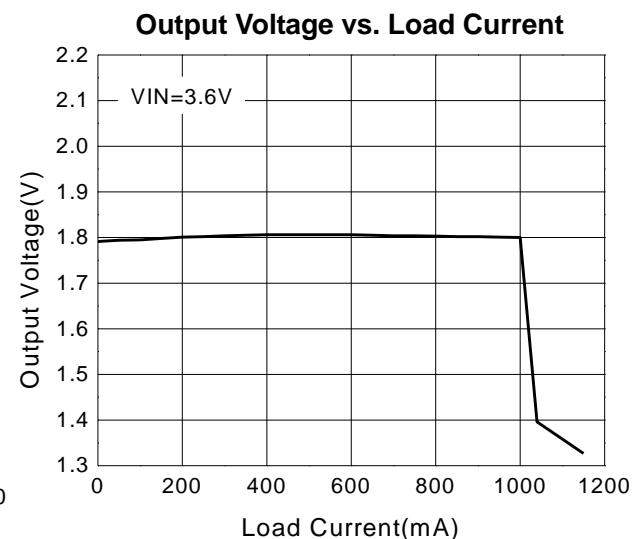
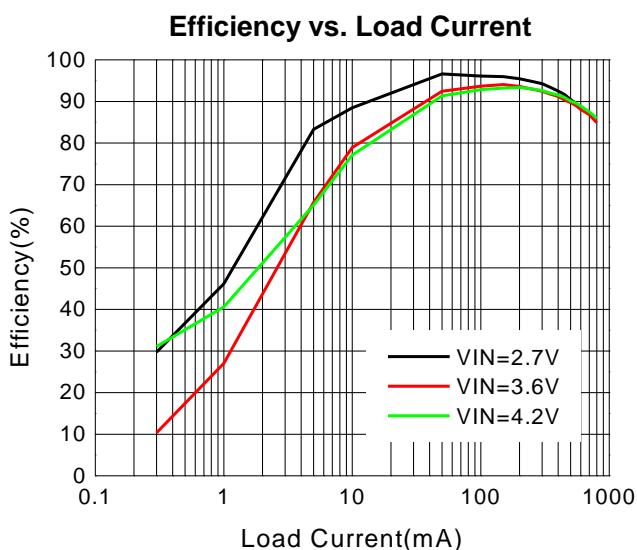
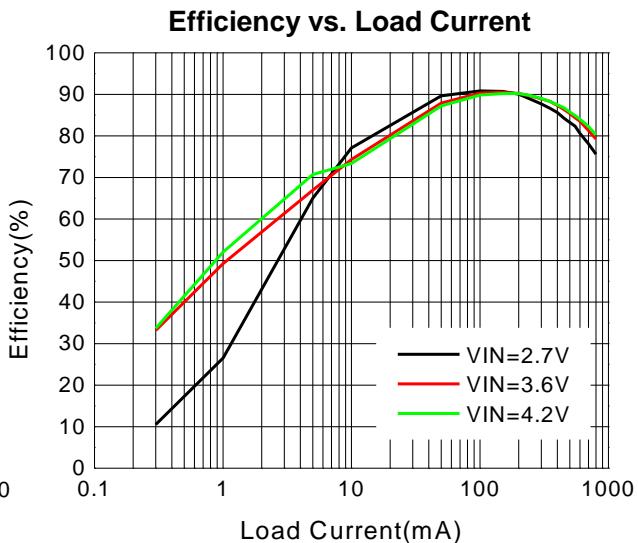
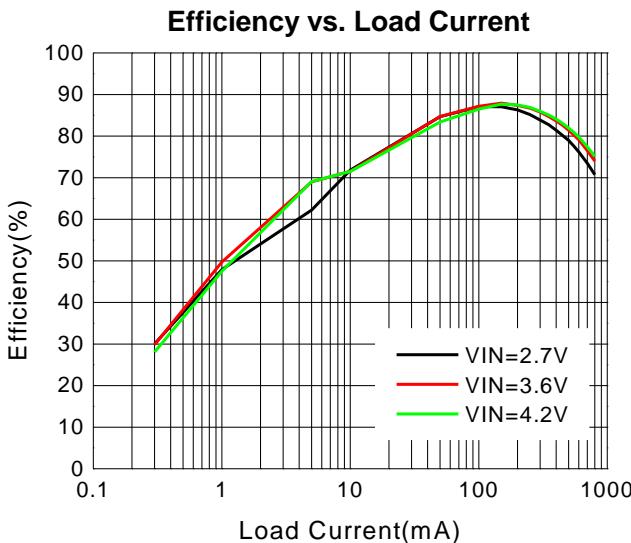
Where T_{ON} is the main switch on time and f_{OSC} is the oscillator frequency (1.5MHz).

Maximum Load Current

The UM3500 will operate with input supply voltage as low as 2.5V, however, the maximum load current decreases at lower input due to large IR drop on the main switch and synchronous rectifier. The slope compensation signal reduces the peak inductor current as a function of the duty cycle to prevent sub-harmonic oscillations at duty cycles greater than 50%. Conversely the current limit increases as the duty cycle decreases.

Typical Performance Characteristics

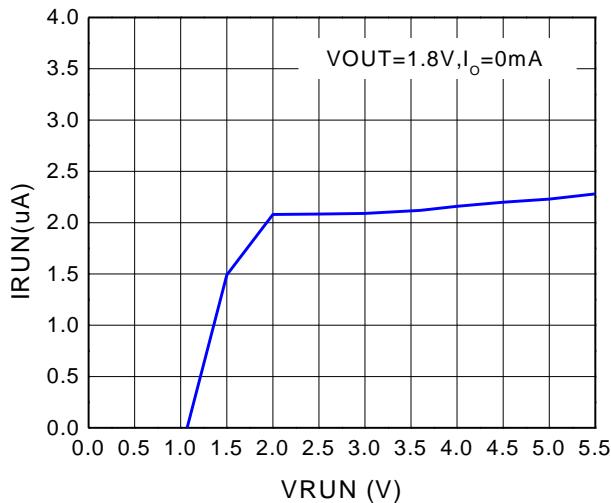
($L_1=2.2\ \mu H$, $C_1=4.7\ \mu F$, $C_3=10\ \mu F$, $T_A=+25\ ^\circ C$, unless otherwise noted.)



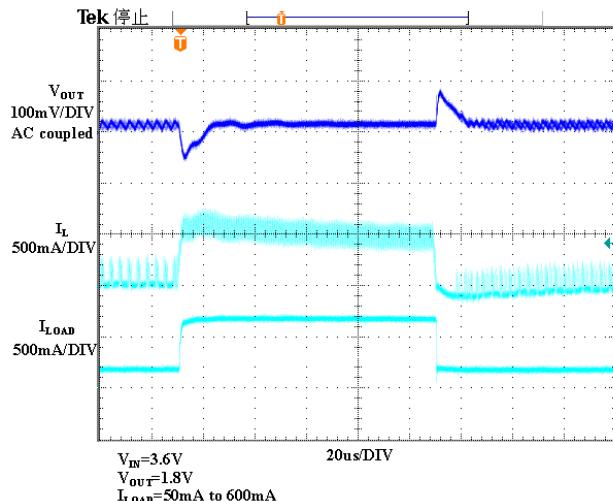
Typical Performance Characteristics (Continued)

(L1=2.2 μ H, C1=4.7 μ F, C3=10 μ F, T_A=+25 °C, unless otherwise noted.)

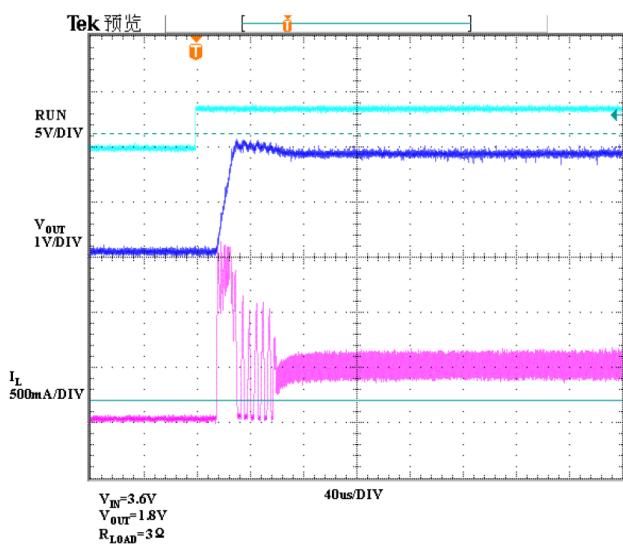
Run Current vs. Run Voltage



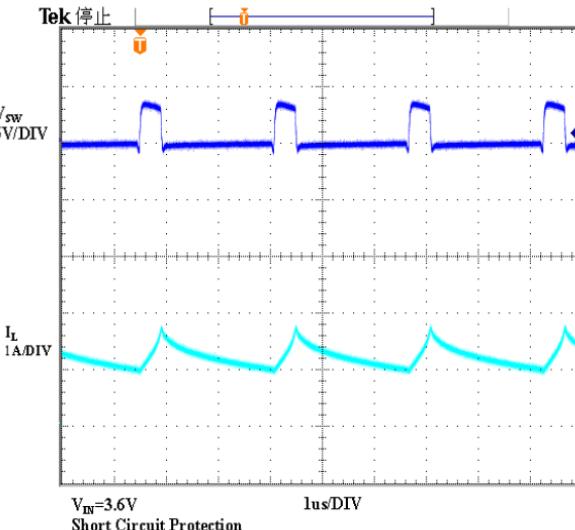
Load Transient



Start-up from Shutdown



Short-Circuit Protection



Applications Information

Output Voltage Setting

The external resistor divider sets the output voltage. The feedback resistor R1 also sets the feedback loop bandwidth with the internal compensation capacitor.

Choose R1 around 300kΩ for optimal transient response. R2 is then given by:

$$R2 = \frac{R1}{\frac{V_{OUT}}{0.6V} - 1}$$

Inductor Selection

A 1 μH to 10 μH inductor with DC current rating at least 25% higher than the maximum load current is recommended for most applications. For best efficiency, the inductor DC resistance shall be <200mΩ.

For most designs, the inductance value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{osc}}$$

Where ΔI_L is the inductor ripple current. Choose inductor ripple current approximately 30% of the maximum load current, 600mA.

The maximum inductor peak current is:

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_L}{2}$$

Under light load conditions below 100mA, larger inductance is recommended for improved efficiency.

Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency shall be less than input source impedance to prevent high frequency switching current passing to the input. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. For most applications, a 4.7 μF capacitor is sufficient.

Output Capacitor Selection

The output capacitor keeps output voltage ripple small and ensures regulation loop stable. The output capacitor impedance shall be low at the switching frequency. Ceramic capacitor with X5R or X7R dielectrics are recommended. The output ripple ΔV_{OUT} is approximately:

$$\Delta V_{OUT} \leq \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{osc} \times L} \times \left(ESR + \frac{1}{8 \times f_{osc} \times C3} \right)$$

Layout Guidance

When laying out the PC board, the following suggestions should be taken to ensure proper operation of the UM3500.

1. The power traces, including the GND trace, the SW trace and the VIN trace should be kept short, direct and wide to allow large current flow. Put enough multiply-layer pads when they need to change the trace layer.
2. Connect the input capacitor C1 to the VIN pin as closely as possible to get good power filter effect.
3. Keep the switching node, SW, away from the sensitive FB node.
4. Do not trace signal line under inductor.

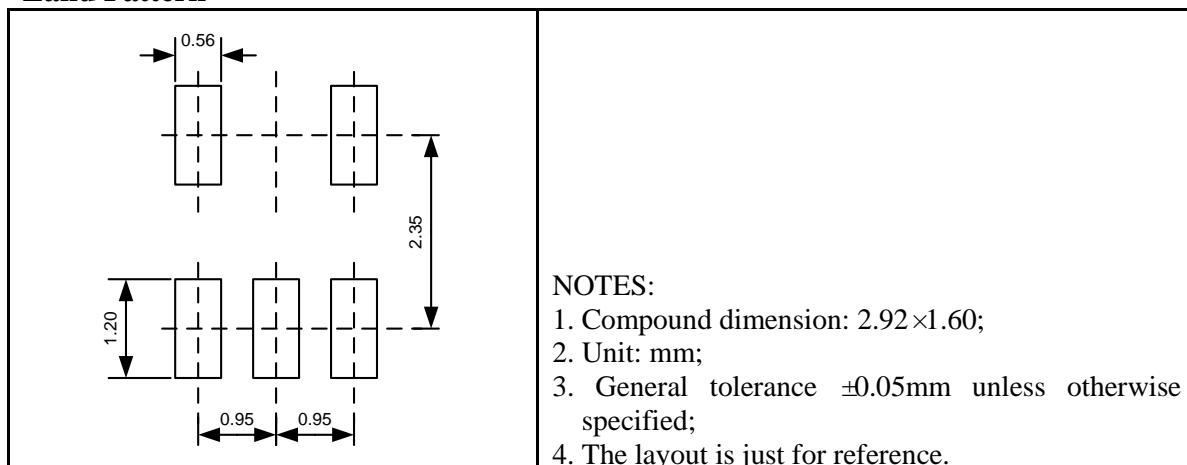
Package Information

UM3500: SOT23-5

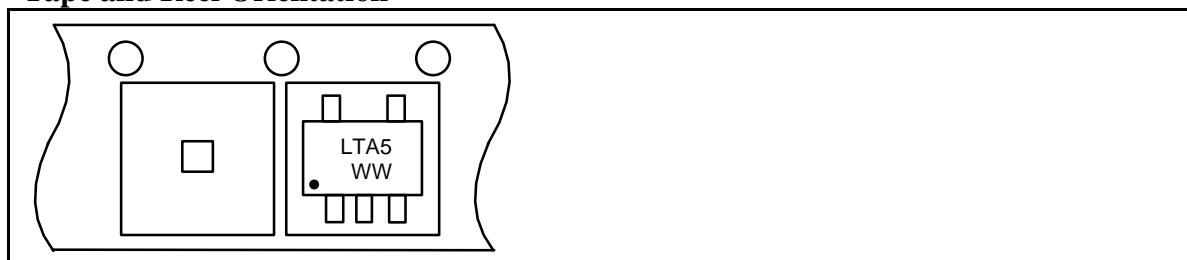
Outline Drawing

| Symbol | DIMENSIONS | | | INCHES | | |
|--------|------------|------|------|----------|-------|-------|
| | Min | Typ | Max | Min | Typ | Max |
| A | 1.013 | 1.15 | 1.40 | 0.040 | 0.045 | 0.055 |
| A1 | 0.00 | 0.05 | 0.10 | 0.000 | 0.002 | 0.004 |
| A2 | 1.00 | 1.10 | 1.30 | 0.039 | 0.043 | 0.051 |
| b | 0.30 | - | 0.50 | 0.012 | - | 0.020 |
| c | 0.10 | 0.15 | 0.20 | 0.004 | 0.006 | 0.008 |
| D | 2.82 | - | 3.10 | 0.111 | - | 0.122 |
| E | 1.50 | 1.60 | 1.70 | 0.059 | 0.063 | 0.067 |
| E1 | 2.60 | 2.80 | 3.00 | 0.102 | 0.110 | 0.118 |
| e | 0.95REF | | | 0.037REF | | |
| e1 | 1.90REF | | | 0.075REF | | |
| L | 0.30 | - | 0.60 | 0.012 | - | 0.024 |
| θ | 0 ° | - | 8 ° | 0 ° | - | 8 ° |

Land Pattern



Tape and Reel Orientation



GREEN COMPLIANCE

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All Union components are compliant with the RoHS directive, which helps to support customers in their compliance with environmental directives. For more green compliance information, please visit:

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