

# 带PWM和单线接口调光功能的36V升压型LED驱动器 UM1663S SOT23-6

## 描述

UM1663S是一款高度集成的LED驱动器IC,能够驱动10个串联WLED。该器件具备一个电流源模式升压转换器,内部集成了一个38V/1A电源开关。

UM1663S支持2.8V至5.5V的宽输入电压范围,并以1MHz的固定频率运行。LED电流通过外部电阻设定,反馈电压输出200mV。工作期间,LED电流由SHDN输入信号的占空比和反馈基准电压决定。

UM1663S支持64级脉冲调光。更多细节见时序图部分。

UM1663S还具有内部过压保护,可防止LED开路时输出超过绝对最大额定值。它采用绿色环保的SOT23-6封装,工作环境温度范围为-40 $\mathbb{C}$ 至+85 $\mathbb{C}$ 。

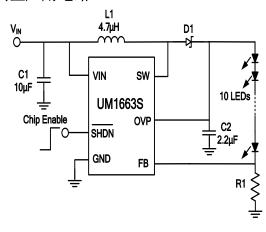
# 应用

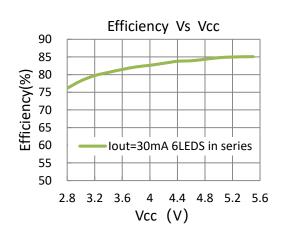
- 移动电话和智能手机
- 数码相机和 GPS
- 便携式 DVD 播放器

# 特性

- 高输出电压: 高达 36V
- 内部软启动和补偿
- 200mV 基准电压
- PWM 调光,频率范围为 500Hz 至 50kHz
- 64 步脉冲调光
- LED 开路保护
- 内部过压保护设定
- 过温保护
- ▶ 限流保护
- 符合 RoHS 标准且不含卤素

## 典型应用电路

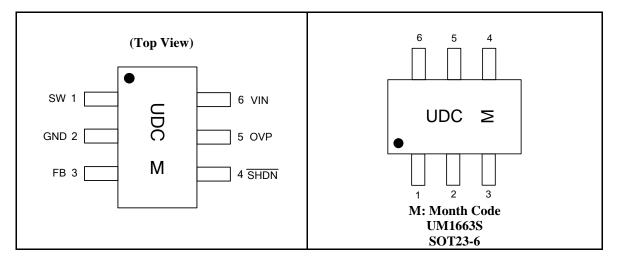






# **Pin Configurations**

# **Top View**



# **Ordering Information**

Part Number	Packaging Type	Marking Code	Shipping Qty
UM1663S	SOT23-6	UDC	3000pcs/7Inch Tape & Reel

# **Pin Description**

Pin Number	Symbol	Function
1	SW	Switch pin. Connect the inductor and the Schottky diode to this pin.
2	GND/PGND	Ground.
3	FB	Feedback voltage pin.
4	SHDN	Chip enable and PWM/Pulse mode dimming pin
5	OVP	Overvoltage protection pin. The pin is connected to an output capacitor to determine overvoltage of the chip. The default overvoltage threshold is 38V.
6	VIN	Supply voltage pin.



# **Absolute Maximum Ratings**

Over operating free-air temperature (unless otherwise noted) (Note 1)

Symbol	Parameter	Value	Unit	
$V_{\rm IN}$	Supply Voltage on VIN (Note 2)	-0.3 to +6	V	
$V_{FB,}V_{/SHDN}$	Voltages on FB, SHDN(Note 2)	-0.3 to +6	V	
$V_{SW}, V_{OVP}$	Voltages on SW, OVP (Note 2)	-0.3 to +42	V	
$P_D$	Power Dissipation at $T_A = 25  ^{\circ}\mathbb{C}$	0.684	W	
	Power Dissipation at $T_A = 70  \text{C}$	0.447		
$\theta_{\mathrm{JA}}$	Package Thermal Resistance(Note 3)	190	°C/W	
$T_{J}$	Operating Junction Temperature	+155	${\mathbb C}$	
$T_{STG}$	Storage Temperature Range	-65 to +150	${\mathbb C}$	
$T_{ m L}$	Maximum Lead Temperature for Soldering 10seconds	+260	${\mathcal C}$	

Note1: Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

Note2: All voltage values are with respect to network ground terminal.

Note3:Junction to Ambient thermal Resistance is highly dependent on PCB layout.

Note4:  $\theta_{JA}$  is measured in the convection at  $T_A$ =25°C (or  $T_A$ =70°C) on a High effective thermal conductivity test board of JESD51-7 thermal measurement standard

Note5: The maximum recommended junction temperature  $(T_J)$  of the UM1663S is 155 °C, the thermal resistance of the UM1663S is  $\theta_{JA}$ =190 °C/W, specified regulator operation is assured to a maximum ambient temperature  $T_A$  of 25 °C .there for the maximum power dissipation is calculated as below:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA} = (155 - 25) / 190 = 0.684W$$

Note 6: The device is not guaranteed to function properly beyond absolute maximum ratings.



# **Electrical Characteristics**

 $(V_{IN}=3.7V, \overline{SHDN}=VIN, T_A=-40 \text{ Cto } 85 \text{ C}, \text{ typical value are } atT_A=25 \text{ C}, \text{unlessotherwise noted})$ 

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
$V_{\rm IN}$	Input Voltage Range		2.8		5.5	V
$I_{\mathrm{Q}}$	Operating	Not switching, V <sub>FB</sub> =0.3V		600		μΑ
	Quiescent Current	Switching, V <sub>FB</sub> =0.1V		1.2	3	mA
$I_{\mathrm{SD}}$	Shutdown Current	SHDN=GND		0.5	2	μΑ
$V_{\rm UVLO}$	Under-voltage Lockout Threshold		2	2.2	2.45	V
	UVLO Hysteresis			0.1		V
$V_{\mathrm{IH}}$	SHDN High Level Input Voltage	VCC=3.7V VCC=5V	1.5			V
V <sub>IL</sub>	SHDN Low Level Input Voltage	VCC=3.7V VCC=5V	1.0		0.6	V
$I_L$	SHDN Input Leakage Current	SHDN=GND or VIN		0.1	1	μΑ
	SHDNDimming Frequency		500		50K	Hz
	SHDNShut Down Delay Time		20			ms
	SHDNPulse	Logic High	0.5			μs
	Dimming Time	Logic Low	0.5			μs
$V_{\mathrm{SW}}$	Maximum Switch Voltage			40		V
$f_{OSC}$	OSC Frequency			1		MHz
$D_{MAX}$	Maximum Duty			93		%
R <sub>DS(ON)</sub>	MOSFET On-Resistance			550		mΩ
	MOSFET Leakage Current	$V_{SW} = 38V$		1	10	μΑ
$I_{LIM}$	MOSFET Current Limit		0.8	1	1.2	A
$V_{OUT}$	Adjustable Output Voltage Range		$V_{\mathrm{IN}}$		36	V
$V_{ m REF}$	Internal Voltage Reference	2.5V≤V <sub>IN</sub> ≤5.5V	190	200	210	mV
	Over Voltage Protection			36		V
	Over Temp Protection			155		$^{\circ}\!$
	OTP Hysteresis			25		$^{\circ}$



ESD AND LATCH UP PERFORMANCE					
I/O Pin ESD-Protection Voltage	Human Body Model		±2		KV
LatchUpPerform ance	JEDEC Standard No.78E		±200		mA

# **Pulse/PWM Dimming Timing Diagram**

For the UM1663S, dimming method is determined by the length of the initial pulse period. If initial pulse width≥2ms, the UM1663S will be set to PWM dimming. If initial pulse width≤1.5ms, then UM1663S will be set to Pulse dimming.

For pulse dimming logic high time  $T_{IH}$  of  $\overline{SHDN}$  should > 0.5us, logic low time  $T_{IL}$  should be between 0.5us  $\sim$  500us.

If  $\overline{SHDN}$  is logic low > 20ms, both in Pulse dimming and PWM dimming mode UM1663S is shutdown.

## **PWM Dimming**

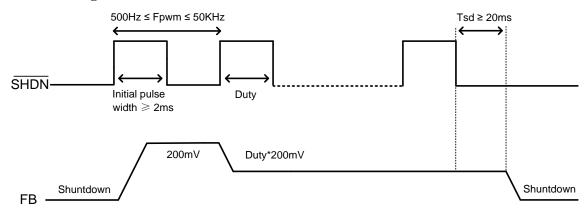


Figure 1. UM1663S PWM dimming

## **Pulse Dimming**

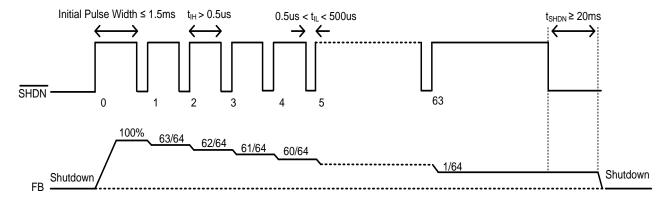
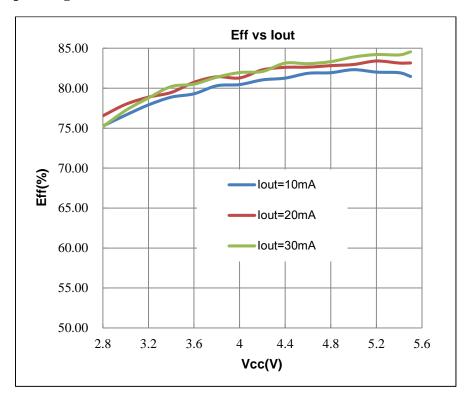
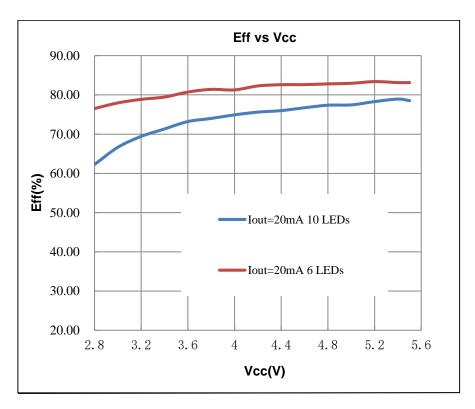


Figure 2. UM1663S pulse dimming



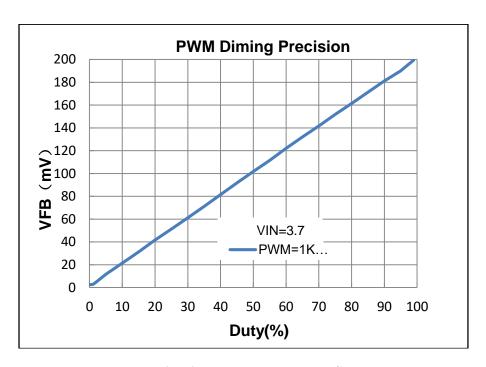


Efficiency Waveform (Vcc=3.6V 6 WLED String)

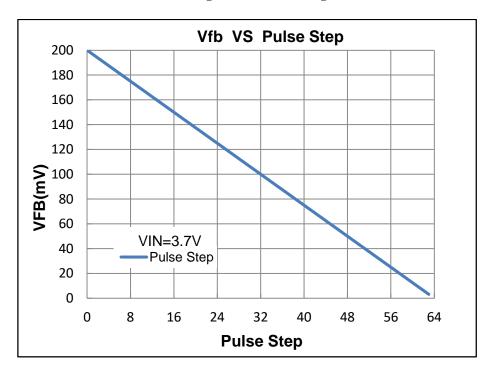


**Efficiency Waveform** 



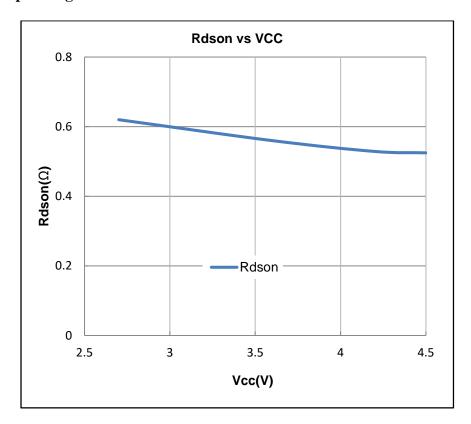


PWM Dimming Pulse vs  $V_{FB}$  Voltage Curve

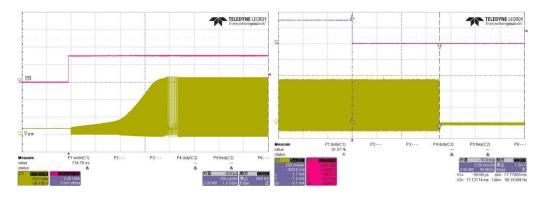


Pulse Dimming Pulse Step vs V<sub>FB</sub> Voltage Curve





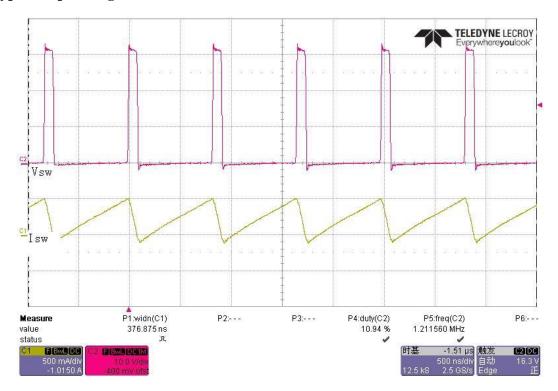
Internal Power NMOS Rds(on) Vs VCC Curve



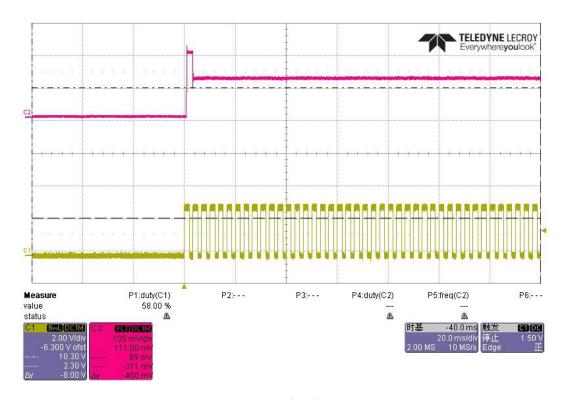
**Power ON from EN** 

**Power OFF from EN** 



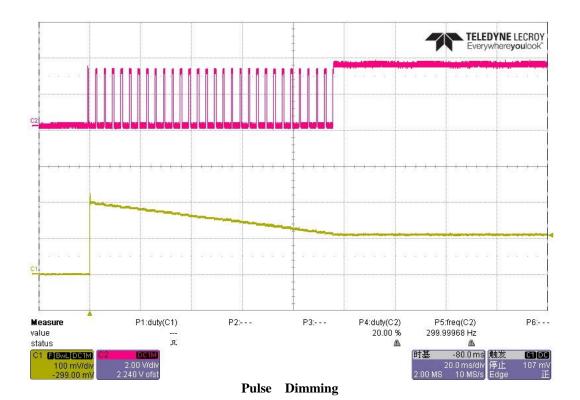


Vsw&IswWaveform(VCC=3.6V Iout=20mA 10 WLED String)



PWM Dimming





# **Function Block Diagram**

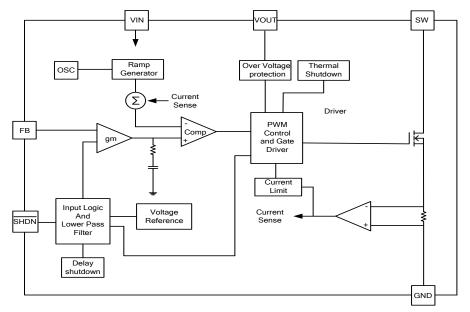


Figure 3. UM1663S function block diagram



# **Functional Description**

The UM1663S uses PWM current-mode boost converter architecture to control the LED current by regulating the feedback voltage. Refer to functional block diagram. The beginning of each cycle turns on the Power MOSFET. A slope compensation ramp is added to the output of the current sense amplifier and the result is fed into the positive input of the comparator (COMP). When this voltage goes above the output voltage of the error amplifier (gM), the Power MOSFET is turned off. The voltage at the output of the gM block amplifies the difference between the reference voltage and the feedback voltage (VFB), so that FB pin voltage can be regulated to the reference voltage.

The UM1663S has built-in soft start to limit the inrush current during startup and to limit the amount of overshoot on the output. Protection features ininclude internal over-voltage protection(OVP), cycle-by-cycle current limit protection and thermal shutdownprotection. OVP protects in the event where an LED fails open, which forces the feedback voltage to zero. This causes the boost converter to operate in maximum duty cycle mode, ramping up the output voltage. Switching will stop when the output reaches the OVP threshold. The OVP feature protects the IC from damaging itself by exceeding the voltage rating on LX pins.

# **Typical Application Circuit**

The UM1663S could drive different WLED topology. For example Figure 4 shows the 3\*7 WLEDS and the current is equal to 90mA. The total WLEDS current could be set by the Rset which is equal to following equation:

#### $I_{total} = 200 \text{mV/R} 1$

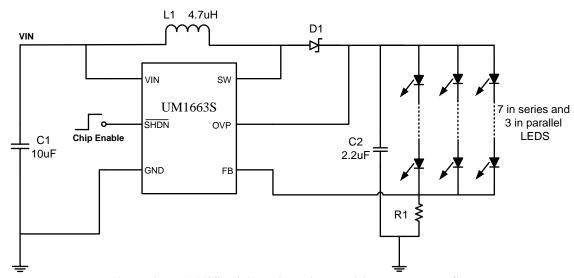


Figure 4. UM1663S driving 7 in series and 3 in parallel LEDS



#### Low input voltage application

In application that have low input voltage range(Figure 5), such as those powered from 1AA cells, the UM1663S may still be used if there is a suitable system supply (such as 3.3V) avaieable to power the controller. In such an application, the inductor maybe connected directly

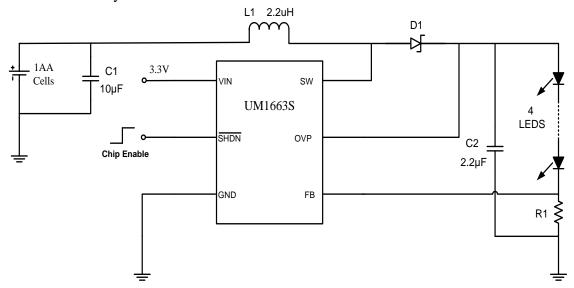


Figure 5. Using UM1663S to drive LED strings power from 1AA Cell

## High input voltage application

In applications that UM1663S have high input voltage range(Figure6), such as those power form a series connected Li-ion batteries. The UM1663S could be used if there is a suitable system supply available (such as 3.3V) to power the controller, or use high input voltage LDO such as UM1460-33 to generate a 3.3V power supply from Li-ion battery. In such cases, the inductor could be connected directly to the battery while the UM1663S power is

Supply by the system supply or output of LDO UM1460-33.

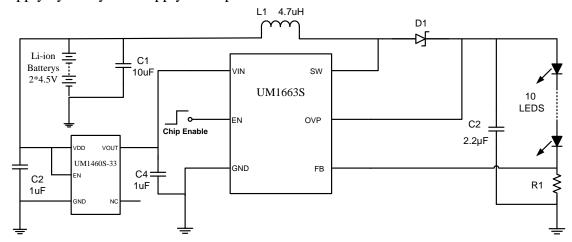


Figure 6. UM1663S based LED driver circuit powered from high input voltage



# **Applications Information**

## **Soft-Start**

The UM1663S includes a soft-start function to avoid high inrush current during start up.

#### **LED Current**

The loop control of the boost converter keeps  $V_{FB}$  equal to a reference voltage,  $V_{REF}$ . Therefore, when RSET is connected between the FB pin and GND, the LED current will be determined by the current through  $R_{SET}$ , which is equal to  $V_{FB}$  /  $R_{SET}$ .

# **Pulse/PWM Dimming**

For the UM1663S, dimming method is determined by the length of the initial pulse period. If initial period ≤1.5ms, the UM1663S will be set to pulse dimming.

If initial period ≥2ms, the UM1663S will be set to PWM dimming. The detail operation of brightness dimming is shown in the Figure 1&Figure 2.

When adding the PWM signal to the  $\overline{SHDN}$  pin, the UM1663S is turned on or off by the PWM signal, so the LEDs operate at either zero or full current. The average LED current increase proportionally with the duty cycle of the PWM signal. The magnitude of the PWM signal should

be higher than the maximum enable voltage of the  $\overline{SHDN}$  pin, in order to let the dimming control perform correctly. The recommended frequency range of the PWM signal is from 500Hz to 50kHz.

Once selection has been made, the dimming method may not be changed until system shutdown, even if the dimming frequency is modified.

#### **Current Limiting**

The current flowing through the inductor during a charging period is detected by a current sensing circuit. If the value is over the current limit, the N-MOSFET will be turned off. The inductor will then be forced to leave charging stage and enter discharging stage. Therefore, the inductor current will not increase over the current limit.

#### **Shutdown Delay**

When the SHDN voltage is in logic low for 20ms (min.) during pulse dimming, the system will enter shutdown.

#### OVP/UVLO/OTP

When the output voltage is higher than a specified value or input voltage is lower than a specified value, the IC will enter protection mode. If the die temperature > 155° C, the IC will also enter protection mode. During protection mode, the P-MOSFET is turned off to prevent abnormal operation.

#### **Inductor Selection**

The recommended Inductor value for 10 WLEDs applications is  $4.7\mu H$ . Small size and better efficiency are the major concerns for portable devices, such as for mobile phone. The inductor should have low core loss at 1MHz and low DCR for better efficiency. The inductor saturation current rating must be greater than the inductor peak current.

#### **Capacitor Selection**

 $10\mu F$  input ceramic capacitor and  $2.2\mu F$  output ceramic capacitor are recommended for driving 10 WLEDs applications. For better voltage filtering, ceramic capacitors with low ESR are recommended. X5R and X7R types are suitable because of their wide voltage and temperature

#### **Diode Selection**

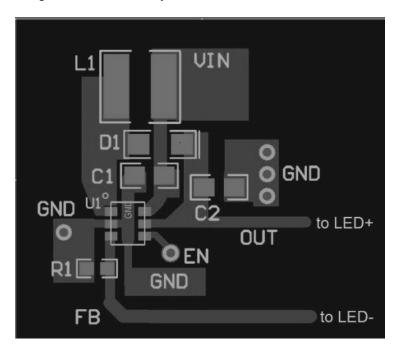
Using a schottky diode is recommended in UM1663 application because of its low forward



voltage drop and fast reverse recovery time. The current raring of the schottky diode should exceed the peak current of the boost converter. The voltage rating should also exceed the target output voltage.

## **Layout Consideration**

PCB layout is very important for high frequency switching regulators in order to keep the loop stable and minimize noise. The input capacitor should be very close to the IC to get the best decoupling. The path of the inductor, schottky diode and output capacitor should be kept as short as possible to minimize noise and ringing. FB is a sensitive node and it should be kept separate from the SW switching node in the PCB layout.



## **Thermal Consideration**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications of the UM1663S, the maximum junction temperature is 155 °C and  $T_A$  is the ambient temperature. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent.

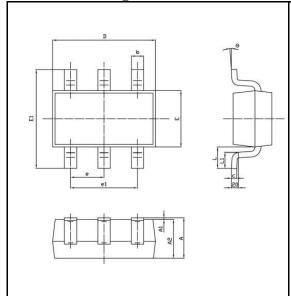
The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ .



# **Package Information**

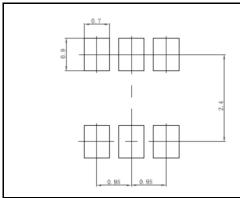
**UM1663S: SOT23-6** 

# **Outline Drawing**



DIMENSIONS					
Symbol	MILLIMETERS		INCHES		
	Min	Max	Min	Max	
A	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
c	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
Е	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950REF		0.037	7REF	
e1	1.800	2.000	0.071	0.079	
L	0.600REF		0.023	BREF	
L1	0.300	0.600	0.012	0.024	
θ	0 °	8°	0 °	8°	

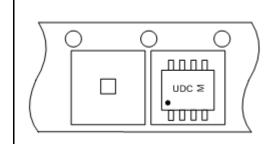
# **Land Pattern**



# NOTES:

- 1. Compound dimension: 2.92×1.60;
- 2. Unit: mm;
- 3. General tolerance ±0.05mm unless otherwise specified;
- 4. The layout is just for reference.

# **Tape and Reel Orientation**





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